

# Microwave Undulator II – Corrugated waveguide

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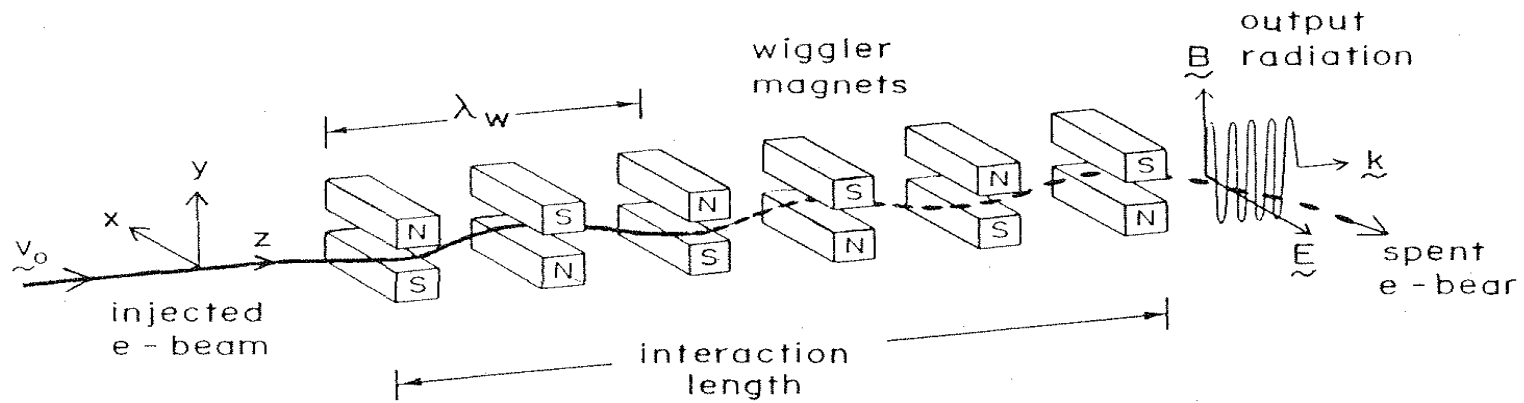
Sami Tantawi



# Acknowledgements

- The idea of using corrugated waveguides for RF undulators was first proposed by Dr. Sami Tantawi
- The work presented in this talk was done at SLAC by me and Dr. Tantawi

# Physics of Undulator



- Highly relativistic electron beam passing through a pump field (wiggler field) produces synchrotron radiation
- Usually constructed from static periodic magnets
- Main undulator parameters are the period and K
  - K typically 2 to 3 for static undulators
  - Typical  $\lambda_u$  for existing static undulators few cm

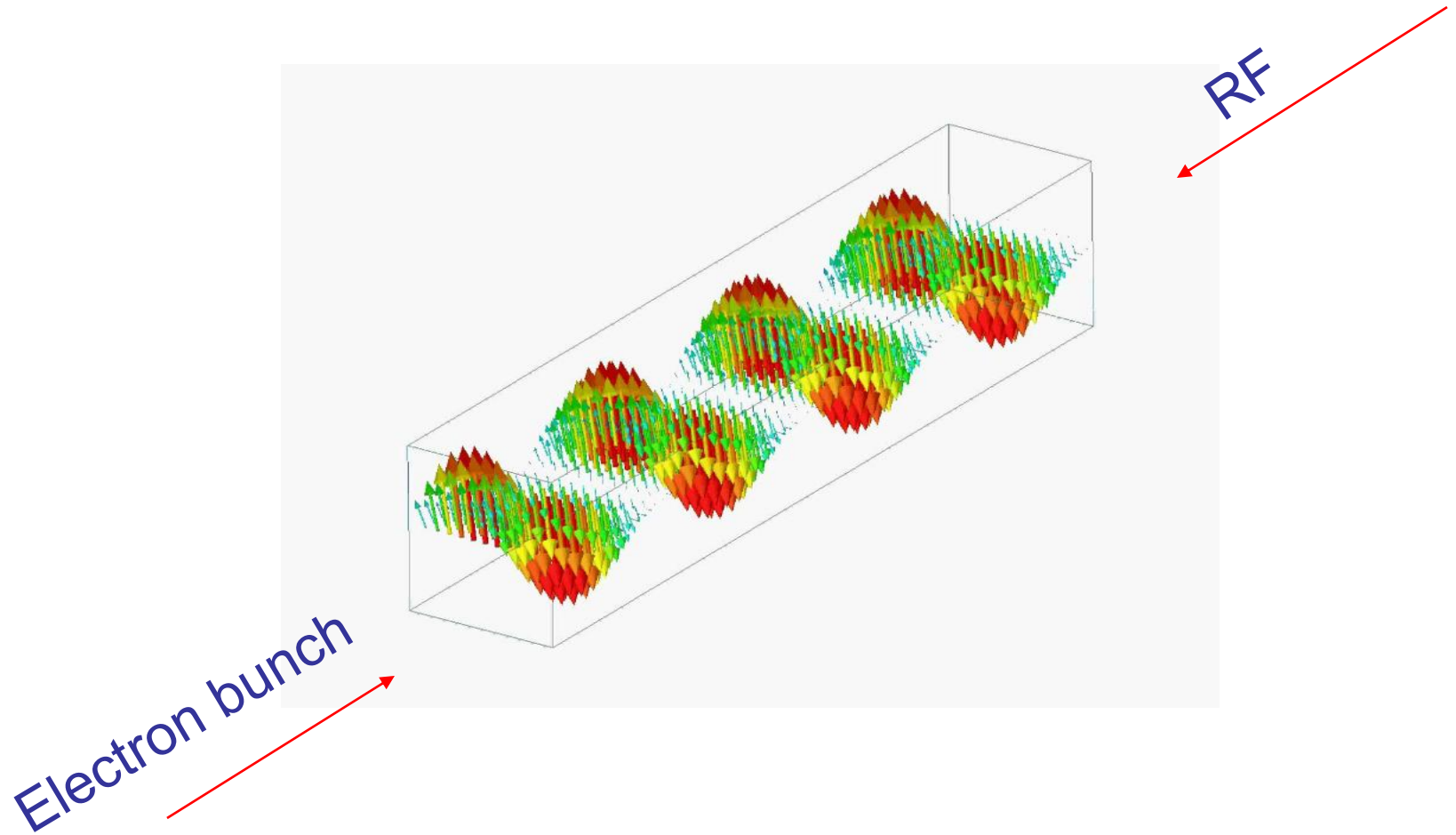
# Why microwave undulator

- Limitations of permanent magnet undulators
  - Polarization cannot be controlled
  - Undulator period cannot be changed
- Advantages of a microwave undulator
  - No magnets to be damaged by radiation
  - Small undulator periods and larger apertures possible
  - Beams with larger radius and emittance can be used
  - Helical undulators are relatively easy with microwaves
- Drawbacks of a microwave undulator
  - High power microwave sources with **precise and stable amplitude and phase** are expensive
  - Handling of tens of GW of microwave power is challenging
  - Design of waveguide/cavity structure can be complicated

# Applications of polarization control

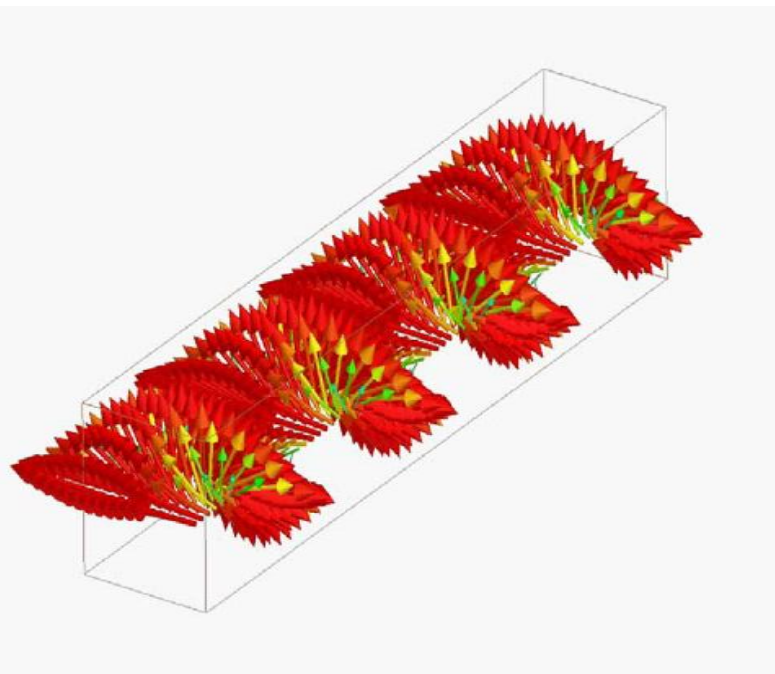
- Exciting scientific opportunity in areas where scattered or absorbed x-ray signal from sample depends on polarization state
- For example, measurement of very small magnetic moment changes in magnetic devices requires fast modulation and lock-in techniques to suppress systemic errors caused by slow drifts which is not possible with pump-probe techniques

# Planar Microwave Undulator

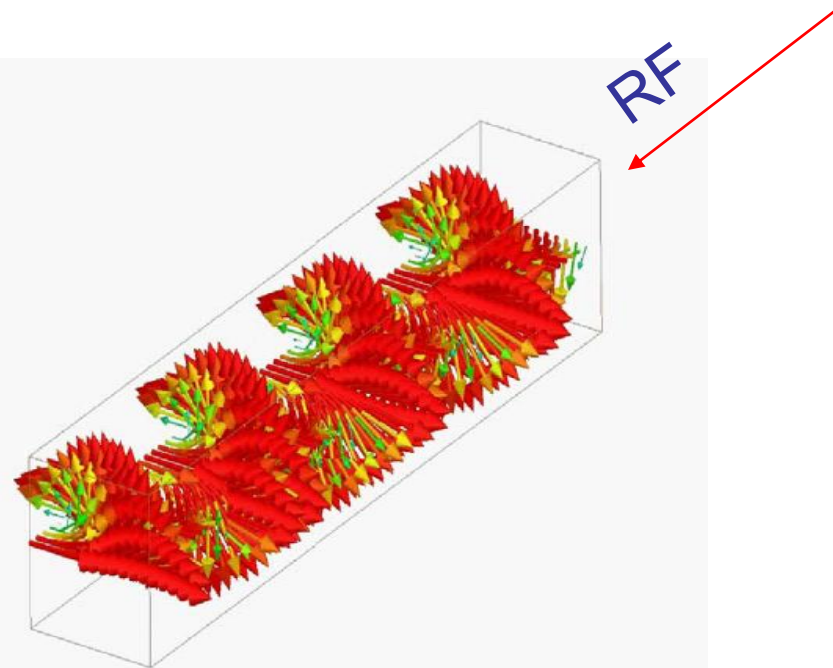


Vector plot of deflecting field in mid-plane of the waveguide

# Helical Microwave Undulator



Right



Left

Electron bunch  
→

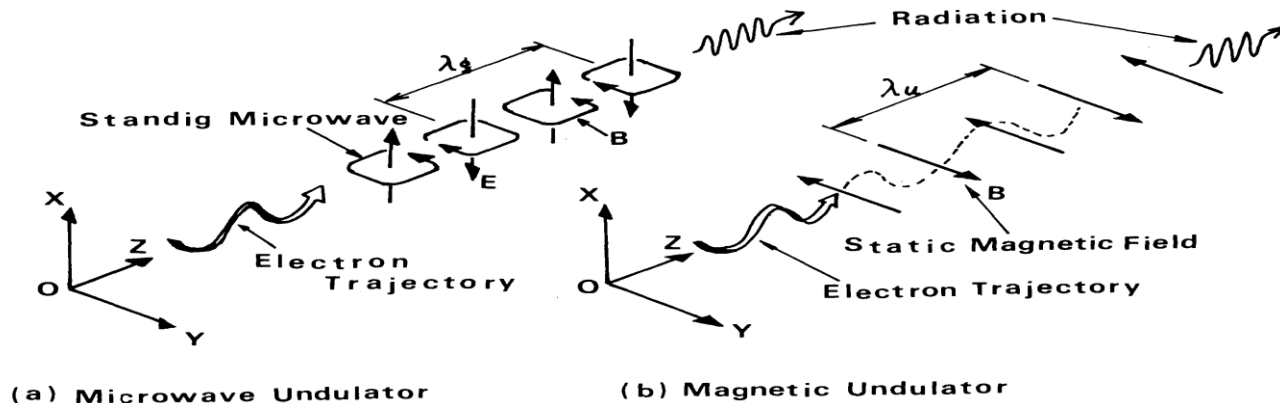
Vector plot of deflecting field in mid-plane of the waveguide

# Microwave undulator realization

- High power handling capabilities of the undulator waveguide have to be maximized and losses in the waveguide minimized
- The operating microwave mode should have maximum field strength in the path of electron beam and minimum tangential RF magnetic field near waveguide wall



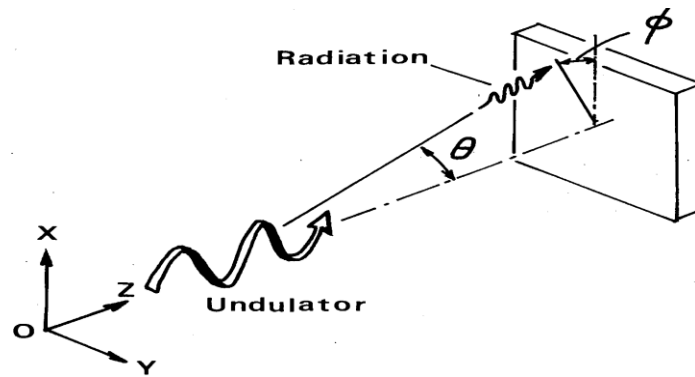
# Structure of microwave undulator



$$\frac{d\beta_x}{dt} = \frac{eB_u}{m_o\gamma} \cos \frac{2\pi z}{\lambda_u} \quad \leftarrow \text{Eqn. of motion for magnetic undulator}$$

$$\frac{d\beta_x}{dt} = -\frac{e}{m_o c \gamma} (E_x - v_z B_y) \quad \leftarrow \text{Eqn. of motion for microwave undulator}$$

# Radiation Characteristics of Undulator



Radiation wave number

Helical Undulator

$$k_s = \frac{2\gamma^2}{1 + 2K^2 + \gamma^2\theta^2} k_u$$

Planar Undulator

$$k_s = \frac{2\gamma^2}{1 + \frac{K^2}{2} + \gamma^2\theta^2} k_u$$

Undulator parameter  $\longrightarrow K = \frac{eB_u\lambda_u}{2\pi m_0c}$

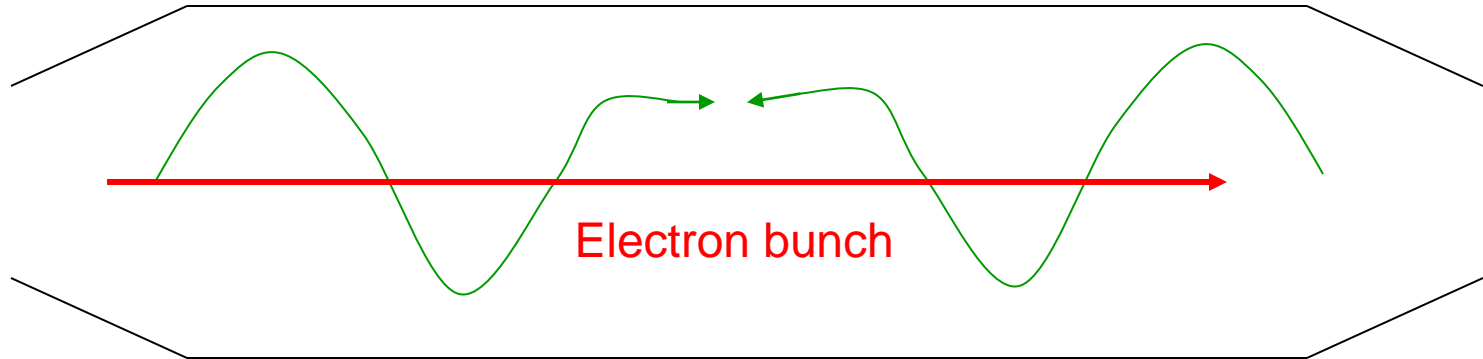
Undulator wave number =  $k_u$

# Need for microwave cavity

Magnetic field in conventional static undulator:  $B_y = B_u \cos \frac{2\pi z}{\lambda_u}$

- To achieve comparable field strength  $B_y$ , tens of GW of RF power is needed.
- RF energy can be stored in a cavity by pumping in RF power to achieve the required field strength  $B_y$
- The required field strength can be sustained by only compensating for RF losses in the cavity

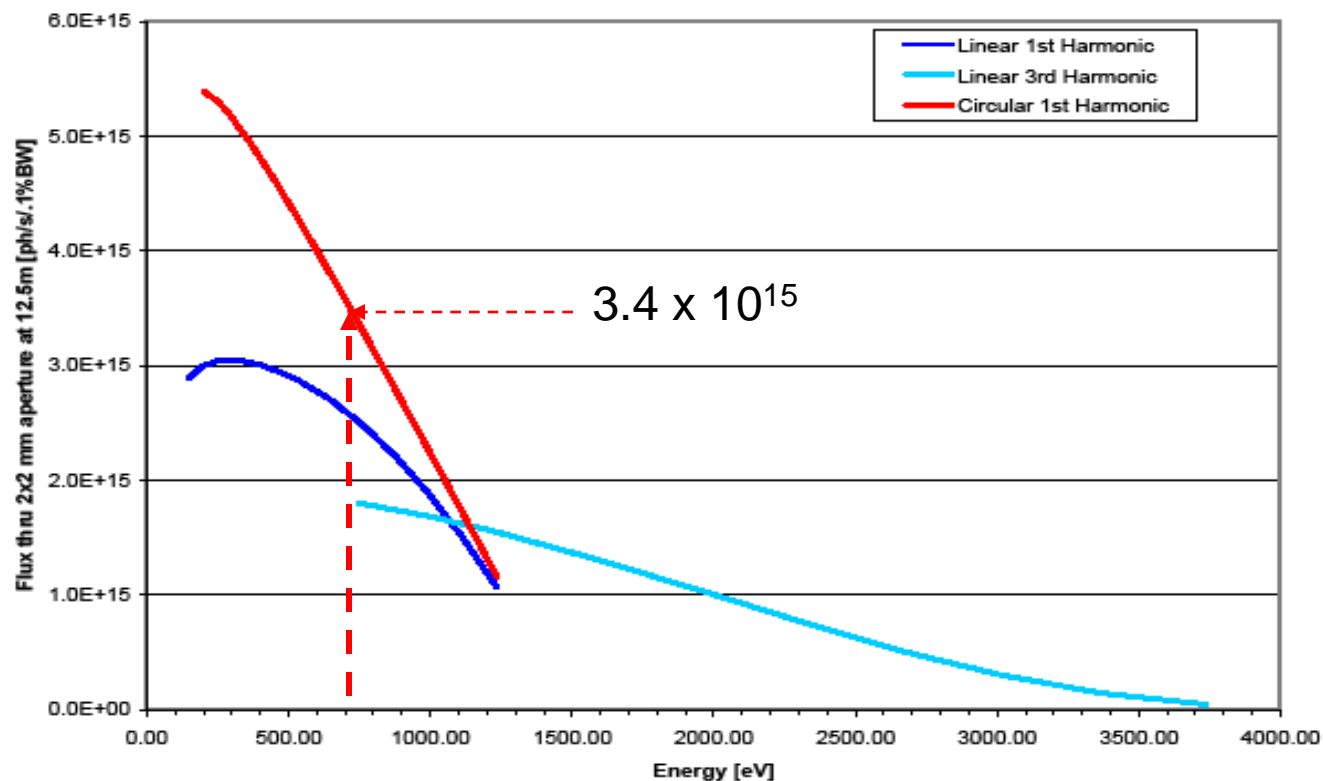
# Theory of microwave undulator



- Electrons can interact with both forward and backward flowing RF waves
- Cavity dimensions are usually large to keep losses low and to reduce sensitivity to dimensions

# Calculated flux curves for Elliptical Polarized Undulator

BL13 EPU, SPEAR 0.1% coupling, 500mA



# of periods = 65,  $\lambda_u = 5.69\text{cm}$ ,  $K=1.07$ , Beam current = 500 mA

# Requirements of SPEAR ring RF undulator to be designed

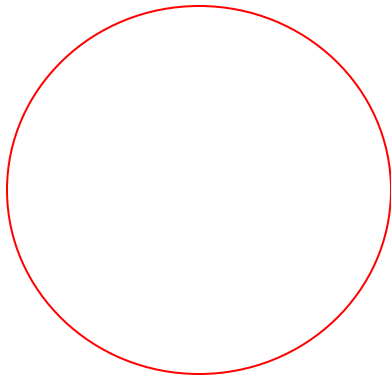
- Electron beam energy: 3 GeV, beam current: 500 mA, aperture: 2 x 2 mm
- Radiation energy: 700-900 eV (circular polarization)
- Fast polarization switch-ability
- Average photon flux and brightness to be **within a factor of 10** compared to static magnetic field Elliptical Polarization Undulator

# Choice of waveguide and modes

- RF field should be “rotate-able” to control radiation polarization
- We consider only circular symmetric waveguide
- Operating RF mode should have strongest field on waveguide axis along path of electron beam
- Very important to keep wall losses extremely low
- Wall losses determine the cost and feasibility of microwave power source

# Waveguide modes considered

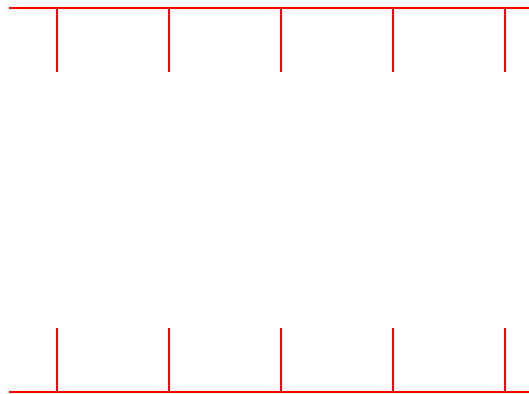
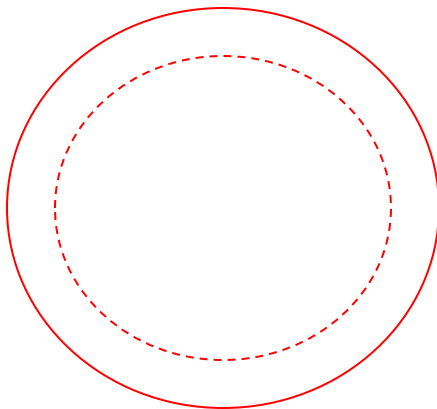
Cylindrical waveguide



$TE_{11}$  mode

$TE_{12}$  mode

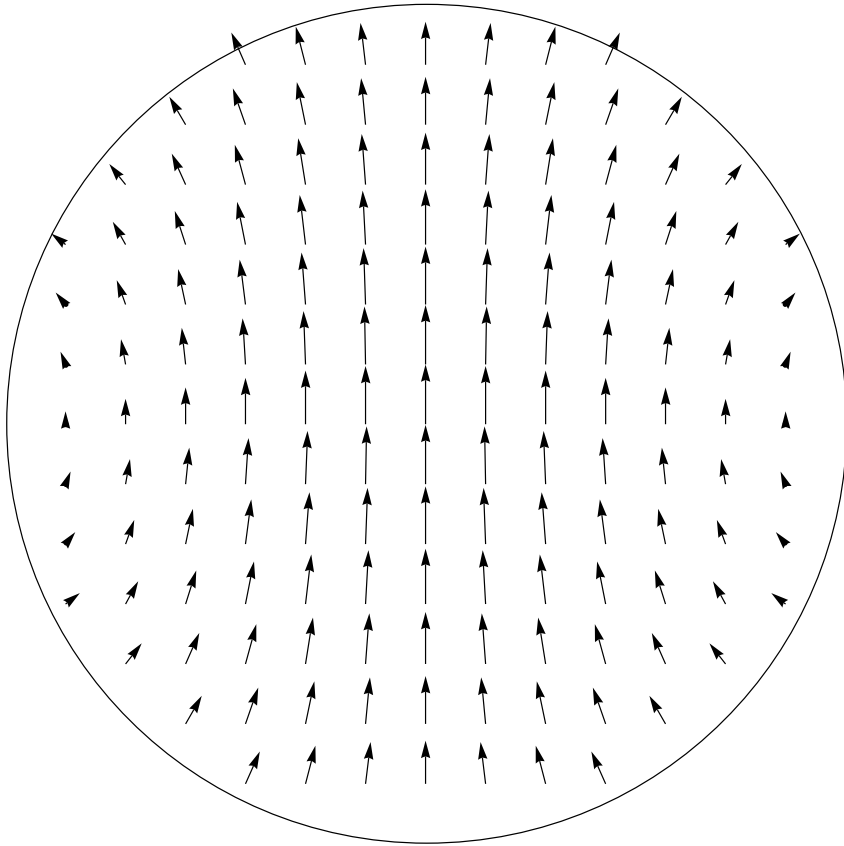
Corrugated waveguide



$HE_{11}$  mode



# Circular waveguide mode $TE_{11}$

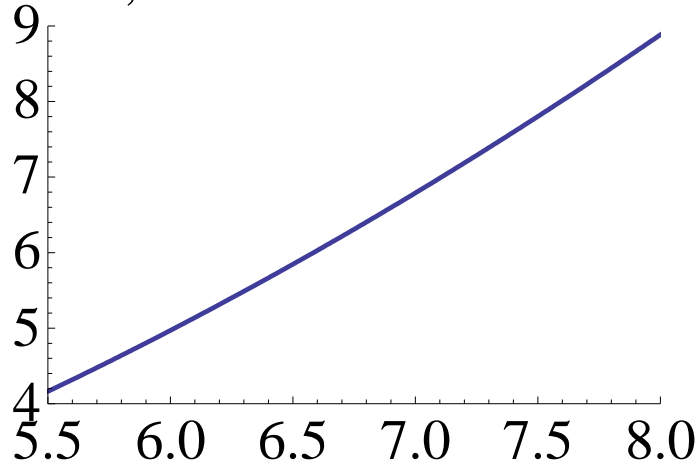


- Fundamental mode, easily excitable
- Has very strong RF field on the axis where the electron bunch travels
- Not the least lossy mode

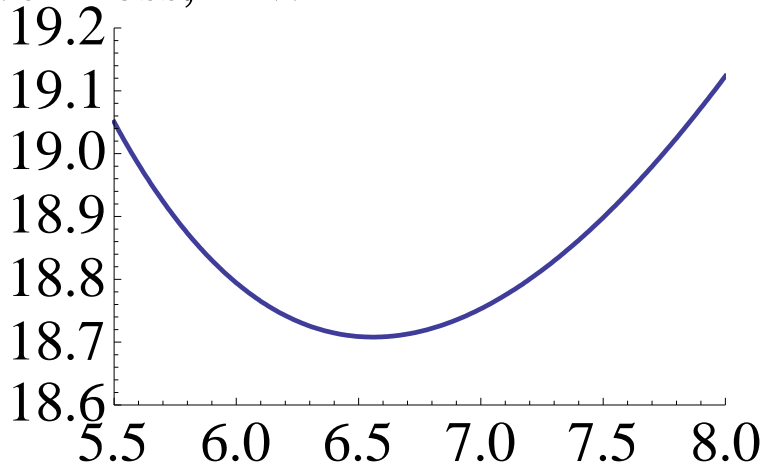
Electric field over waveguide cross-section

# Power loss in a TE<sub>11</sub> circular waveguide undulator

Power Flow, GW



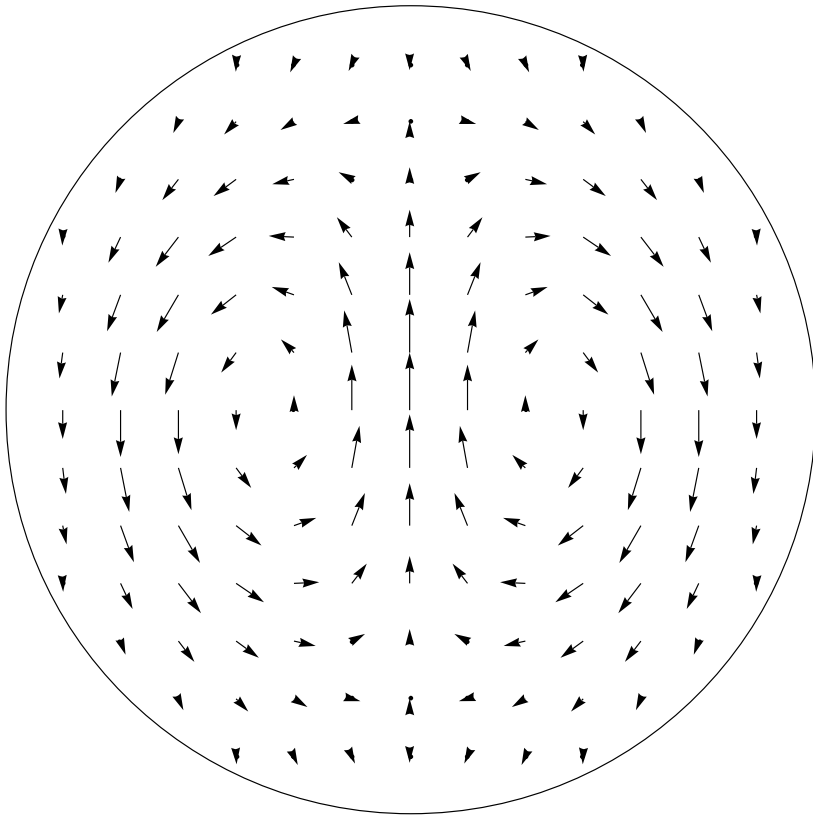
Power Loss, MW



Waveguide radius, cm

- Flux = 1/5th of Static undulator flux ( $3.4 \times 10^{15}$  [ph/s/0.1 % BW])
- $K = 0.71$
- Circular polarization
- Photon energy = 700 eV
- $L = 3.7$  m
- $\lambda u = 6.1$  cm
- $f = 2.6$  GHz

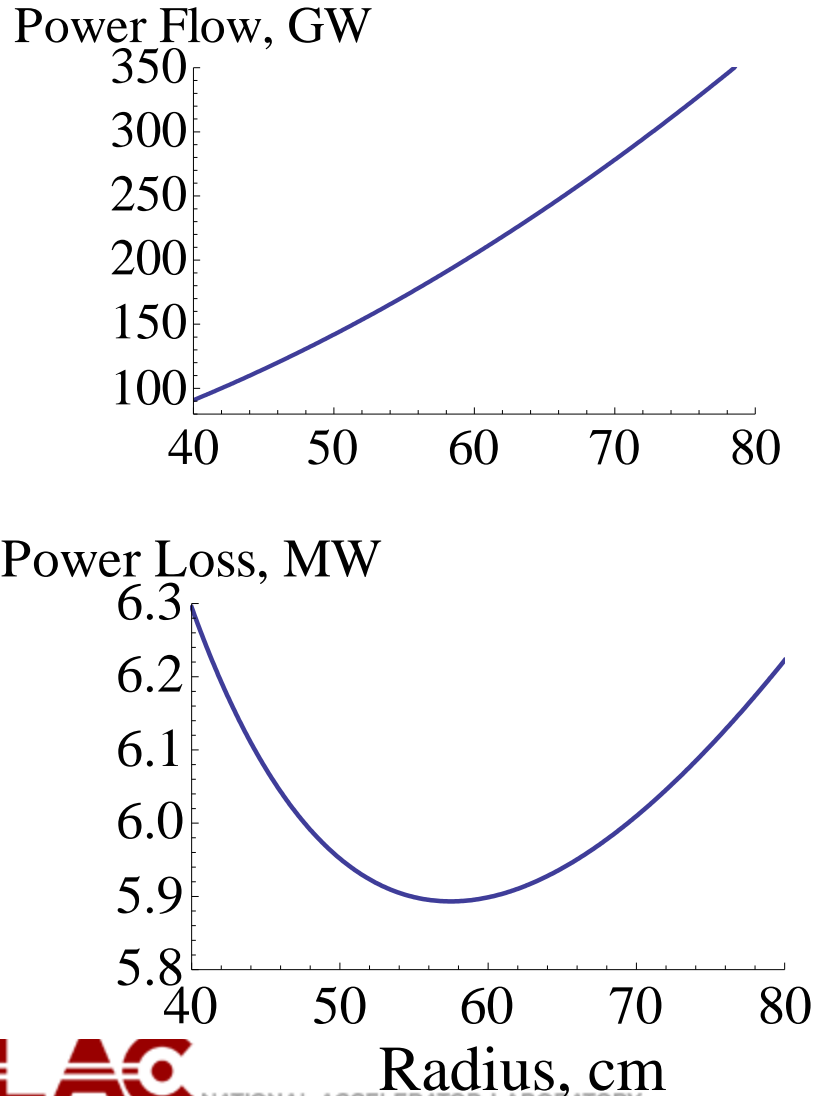
# Circular waveguide mode $TE_{12}$



- Has same field structure at the axis as  $TE_{11}$  mode
- Needs much larger waveguide radius and power
- Attenuation is much lower than  $TE_{11}$  mode

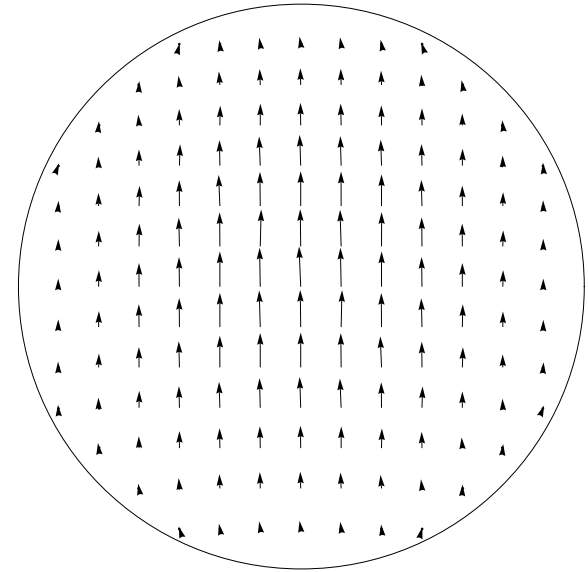
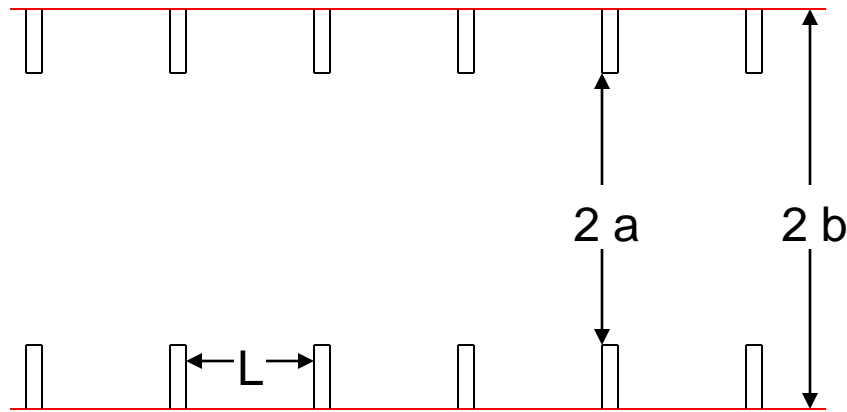
Electric field over waveguide cross-section

# Power loss in a TE<sub>12</sub> circular waveguide undulator

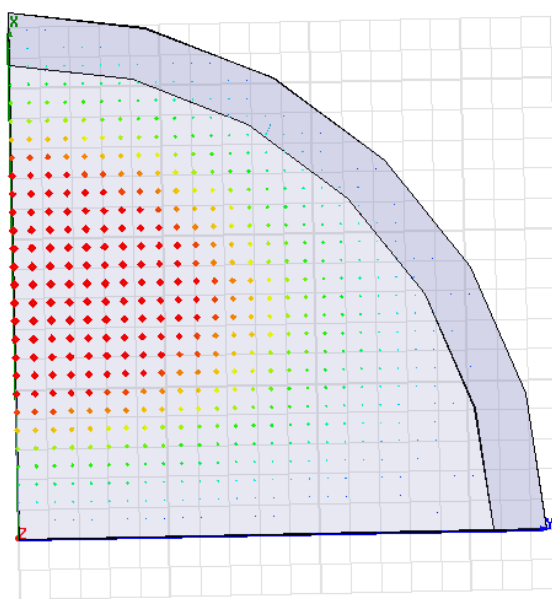


- Flux = 1/5th of Static undulator flux ( $3.4 \cdot 10^{15}$  [ph/s/0.1 % BW])
- $K = 0.68$
- TE<sub>12</sub> mode, Circular polarization,
- Photon energy = 700 eV
- $L = 3.7$  m
- $\lambda_u = 6.35$  cm
- $f = 2.38$  GHz

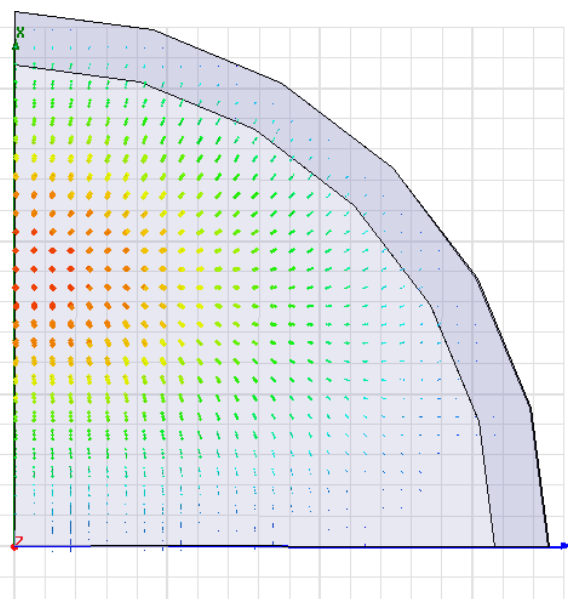
# Corrugated waveguide, hybrid $HE_{11}$ mode



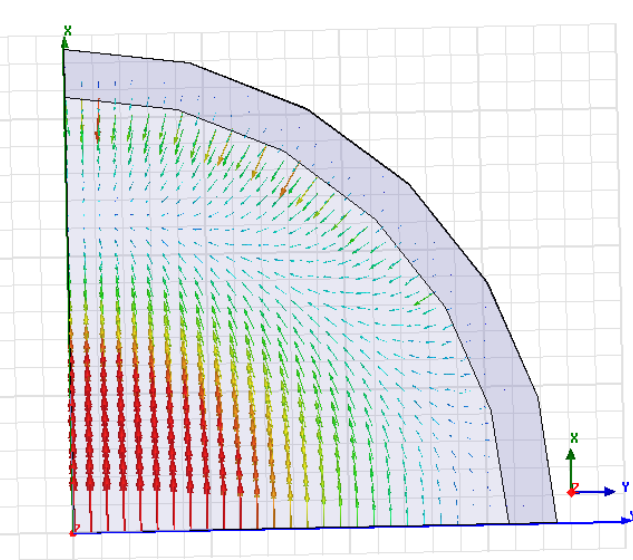
- Combined TE and TM modes lead to hybrid modes
- Under “balanced hybrid” conditions, the field transforms into a low loss linearly polarized wave
- The field is strongly linearly polarized on the axis which is highly desirable for undulator operation
- The field is extremely low near the waveguide walls translating to very low RF losses



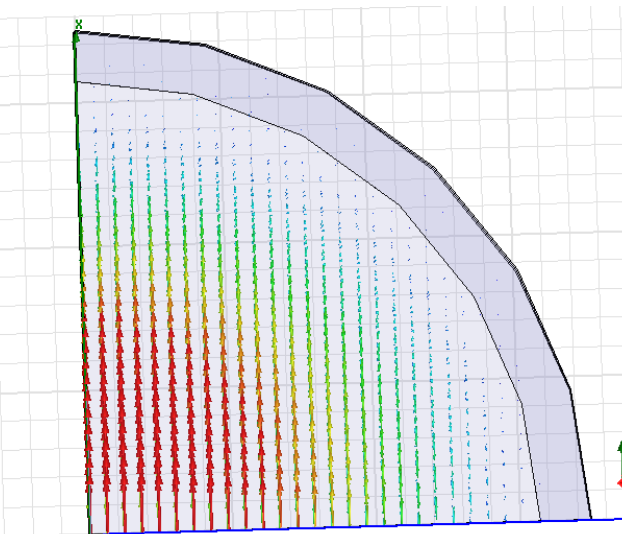
$\beta L = 0.05 \text{ rad}, Q = 36945$



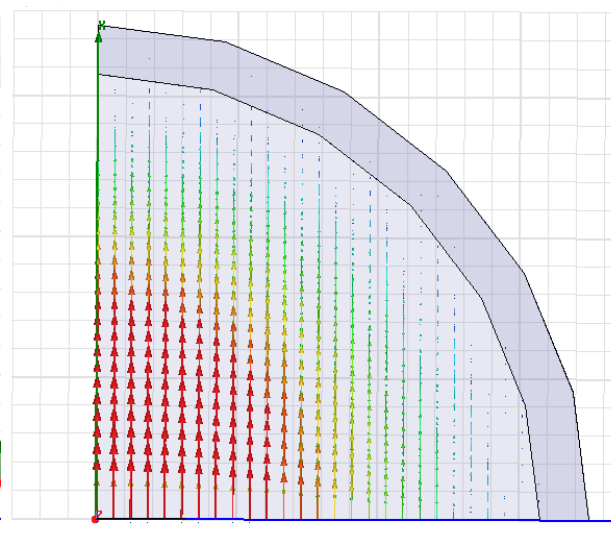
$\beta L = 0.3 \text{ rad}, Q = 42537$



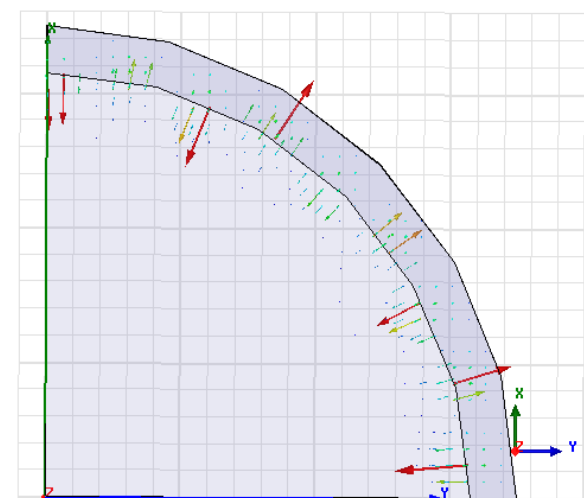
$\beta L = 0.5 \text{ rad}, Q = 74322$



$\beta L = 1 \text{ rad}, Q = 1.05 \times 10^6$



$\beta L = 1.65 \text{ rad}, Q = 5.5 \times 10^6$



$\beta L = 2 \text{ rad}, Q = 13574$

# Analysis of corrugated waveguide (neglecting space harmonics)

$$r < a$$

$$F_z = AJ_m(k_r r)e^{jm\phi}; A_z = AZ_z J_m(k_r r)e^{jm\phi}$$

$$E_z = -j \frac{k_r^2}{k\sqrt{\mu\epsilon}} A_z; H_z = -j \frac{k_r^2}{k\sqrt{\mu\epsilon}} F_z$$

$$E_r = -\frac{1}{\epsilon r} \frac{\partial F_z}{\partial \phi} - \frac{\beta_z}{k\sqrt{\mu\epsilon}} \frac{\partial A_z}{\partial r}$$

$$a < r < b$$

$$A_z = jA \frac{\sqrt{\mu\epsilon}}{k} J_m(k_r a) \frac{H_m^{(2)}(kr)H_m^{(1)}(kb) - H_m^{(1)}(kr)H_m^{(2)}(ka)}{H_m^{(2)}(ka)H_m^{(1)}(kb) - H_m^{(1)}(ka)H_m^{(2)}(kb)} e^{jm\phi}$$

$$E_z = -j \frac{k}{\sqrt{\mu\epsilon}} A_z; H_r = \frac{1}{\mu r} \frac{\partial A_z}{\partial \phi}; H_\phi = -\frac{1}{\mu} \frac{\partial A_z}{\partial r}$$

- Both  $TE_z$  and  $TM_z$  modes are present

- Balanced Hybrid mode is possible only when waveguide impedance  $Z_z \approx$  Free space wave impedance

- Only  $TM_z$  mode present

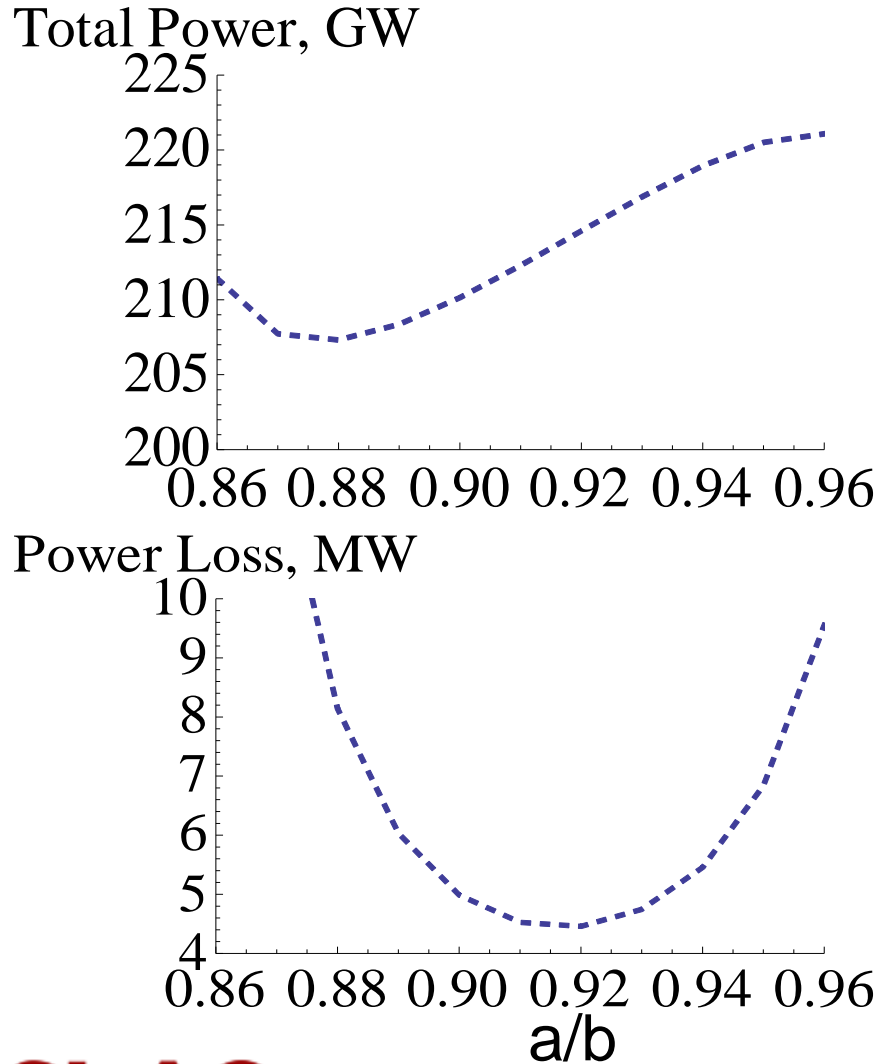
- Balanced Hybrid condition possible when  $(b - a) \approx \lambda/4$

# Boundary conditions for corrugated waveguide

- For sufficiently small slot width, no  $TE_z$  mode present inside corrugation
- Then, at  $r = a$ ,  $E_\phi = 0$
- Admittance  $H_\phi / E_z$  is continuous at  $r = a$
- Equating admittance for the two set of equations at  $r = a$  gives the characteristic dispersion equation for the corrugated waveguide

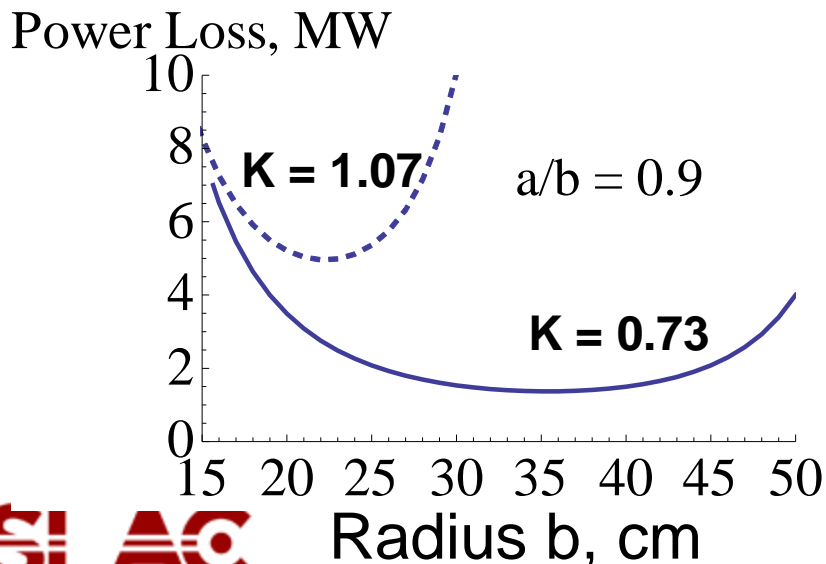
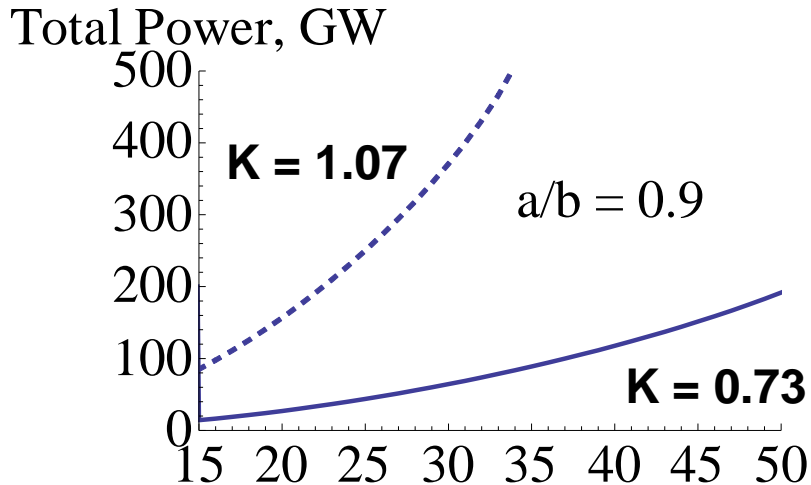


# Power loss dependence on corrugation depth



- Flux = Static undulator flux ( $3.4 \cdot 10^{15}$  [ph/s/0.1 % BW])
- **K=1.07**
- **Photon energy = 700eV**
- **L = 3.7 m**
- **b = 23 cm**
- **f = 4.05 GHz**
- **$\lambda u = 5.92$  cm**

# Power loss in a $HE_{11}$ corrugated waveguide undulator



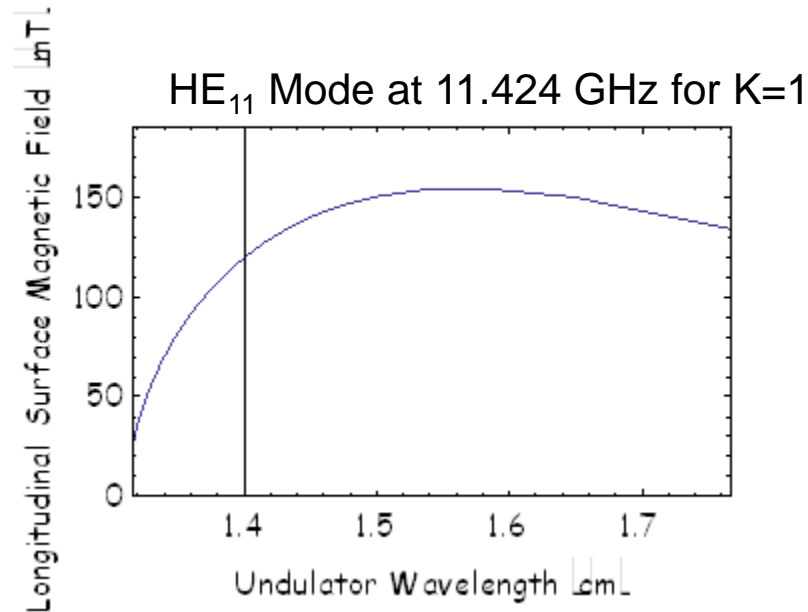
- Flux = Static undulator flux
- $K = 1.07$
- $f = 4.05$  GHz
- $\lambda_u = 3.71$  cm

- Flux = 1/5th Static undulator flux
- $K = 0.68$
- $f = 2.55$  GHz
- $\lambda_u = 5.92$  cm

# Superiority of $HE_{11}$ - mode

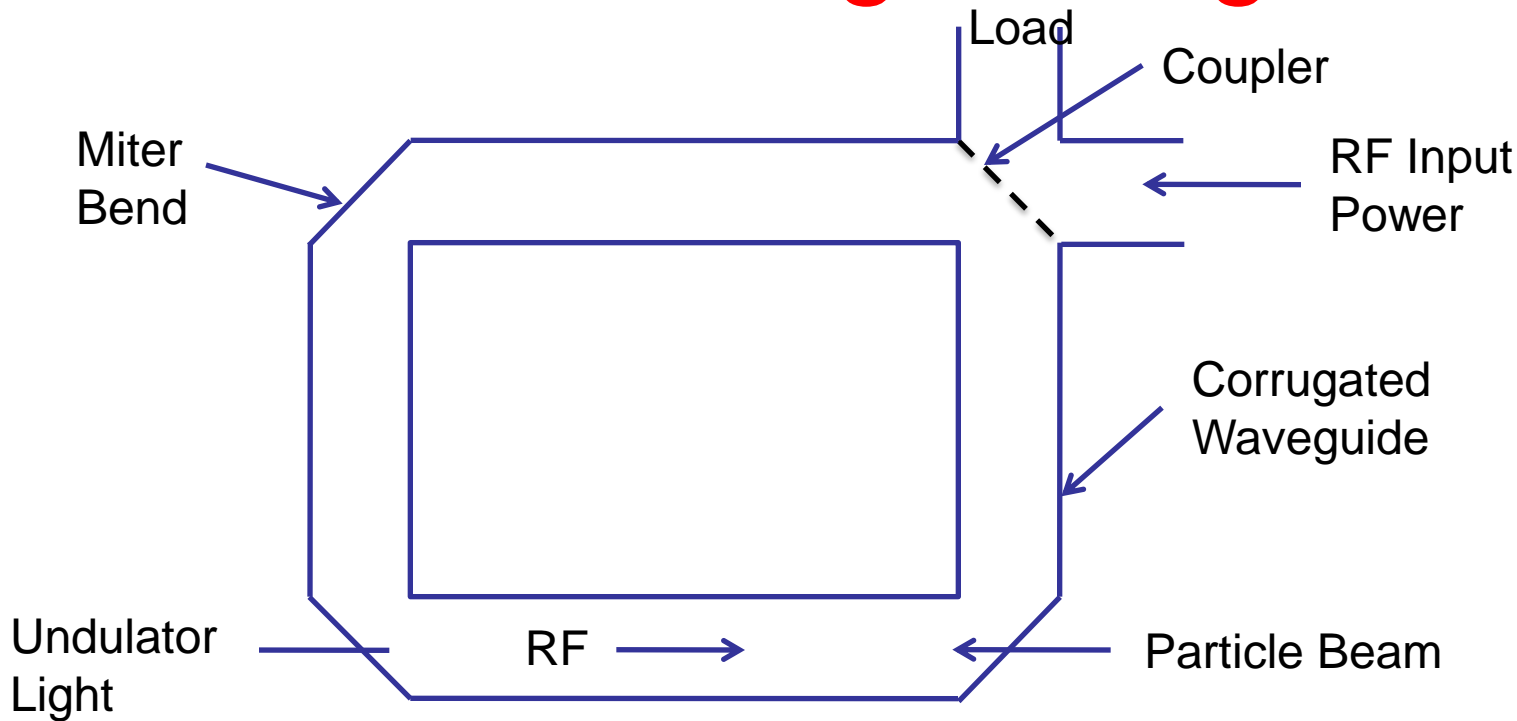
|                       | $TE_{11}$ | $TE_{12}$ | $HE_{11}$ |
|-----------------------|-----------|-----------|-----------|
| Undulator parameter K | 0.71      | 0.68      | 0.68      |
| Power flow (GW)       | 5.8       | 180       | 79        |
| RF power loss (MW/m)  | 5.1       | 1.6       | 0.326     |
| RF frequency (GHz)    | 2.64      | 2.38      | 2.37      |
| Cavity Radius (cm)    | 6.5       | 57.7      | 38        |

# Superconducting RF Undulator



- Surface magnetic field is less than the quenching field of niobium
- Allows application of RF undulator to storage ring applications, where CW or quasi-CW operation are required
- RF power needed is only few hundred watts to kilo watts – sources readily available

# Resonant Ring Configuration



- A closed ring with length  $n\lambda g$
- Tune by adjusting ring length
- Considerable development for relevant components (miter bend, couplers) has been done (ITER transmission lines)

# Conclusions and future work

- Corrugated waveguide is an attractive option for a microwave undulator due to its low losses
- Successful development will enable design of undulators with capabilities not possible with current static undulators
- Could lead to a new class of FEL and storage ring undulators